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Application Of Optimization Technique In Designing Robust PSS For Two Machine System

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ABSTRACT— In this paper, the Power System Stabilizer is designed using LQR and Robust $H\infty$ loop shaping technique in order to maintain the stability of the system. The weighting function of Robust $H\infty$ loop shaping technique is determined using Genetic Algorithm. The main objective of this paper is to achieve desired stability and analyze the robustness of the centralized controller for multi-machine system and the comparison is made between the LQR based PSS and Robust controller. The LQR based PSS is tested even under perturbations. The suitability of the proposed approach for PSS proves the robustness of the proposed design.

KEYWORDS— LQR, Robust $H\infty$ loop shaping, Robust controller, Power System Stabilizer, damping;

NOMENCLATURE

K_s Stabilizer gain

- T_w Time constant of washout filter
- T₁-T₄ Time constant of compensator blocks
- V_{ref} Reference voltage
- H Inertia constant in MW.s/MVA
- D Damping coefficient.
- δ Rotor angle in electrical radian.
- ω Rotor speed in p.u.
- ω_d Damped frequency in Hz
- ς Damping ratio
- $\Delta \omega_{\rm r}$ Deviation in rotor speed p.u
- ΔV_t Incremental change in terminal voltage
- $\Delta V_{ref} \quad \text{Incremental change in reference Voltage}$
- ΔT_{mech} Incremental change in mechanical torque
- K_r LQR controller gain
- Q, R LQR weighting function

I. INTRODUCTION

The main objective of installing power system stabilizer (PSS) is to achieve desired stability and security at a reasonable cost by adding damping to electromechanical oscillations. They were developed to extend stability

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limits by modulating the generator excitation to provide additional damping to the oscillations of synchronous machine rotors. In recent years there has been an increasing interest on applying advanced control designs in power systems like adaptive control, $H\infty$ control, μ synthesis, nonlinear control, feedback linearization, fuzzy logic control and neural control. The goal of these studies is to achieve stability and performance robustness.

In [1] PSS has been developed based on a reasonable area model, which consist of component models, models of component controllers, and area grid models. A new structure for stability-enhancing excitation controllers designed using a nonlinear multi-machine system model and Lyapunov direct method has been described in [2]. An analytical method so called minimum phase control loop and intelligent method using a Genetic Algorithm (GA) has been proposed in [3] for an offline Power System Stabilizer (PSS). The application of wave variable method in reducing the effect of delay during the transmission of control signals via large distances in a multi-machine power system has been introduced in [4]. In [5] a systematic approach to damp the low frequency oscillations is observed in Three Machine Nine Bus Multi-machine Power Systems based on Genetic Algorithm (GA). The Optimal Controller design problem is formulated as an optimization criterion comprising of Time domain based objective function to compute the optimal controller parameters. In [6] the proposed RPSS is free from common deficiencies at normal power system nonlinear controllers which are network dependent and equilibrium dependent. In [7] PSS have been designed using Glover-McFarlane H^{\$\phi\$} loop-shaping technique for multimachine system based on sequential tuning.

In this paper, the PSS has been developed using various methods such as LQR and Robust control design. A new thing in this paper is application of Optimization algorithm (Genetic Algorithm) to determine the weights in designing

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Robust PSS. Generally optimization technique is applied in designing robust PSS to determine the parameters of PSS only, but here it is applied to determine weights in designing Robust PSS. Conventional stabilizers are not designed in a way to guarantee the desired level of robustness. Such designs are specific for a given operating point; they do not guarantee robustness for a wide range of operating conditions. The resulting LQR-PSS and Robust H ∞ loop shaping procedure based PSS ensures the stability for a set of operating points with respect to the nominal system and has good oscillation damping ability. The LQR based PSS is tested even under perturbed condition such as change in mechanical torque.

II. PROBLEM DESCRIPTION

In Fig 1, Generators gen1 and gen2 are local generators and the rest of the system is approximated by infinite bus system. The two generators are of different ratings, one with 920.35MVA, 18KV, 60Hz and the other with 911MVA, 26KV, 60Hz units. In simulation studies, each of these two generators will be represented by a transient model. With an infinite bus in the network, its voltage phasor can be conveniently chosen as the reference phasor for the angle of other bus voltages and rotor angles. If there is no infinite bus, the rotor q-axis of one of the generators may be chosen as the reference axis instead. In a simulation, the rotor speed of the reference machine can be held steady by setting its inertia to some very large value. The PSS is designed for two generator infinite bus system using various techniques such as LQR and GA tuned robust controller.



Fig. 1. Two-machine four-bus system

A. Modeling of Two generator infinite bus system

The dynamic equation of the rotor, in per unit, using the relationship between rotor angle, δ , and the rotor speed, ω_r , as

$$\frac{1}{\omega_{\rm b}}\frac{\mathrm{d}^2\delta}{\mathrm{dt}^2} = \frac{1}{2\mathrm{H}}\left(\mathrm{T}_{\mathrm{mec}\,\mathrm{h}} + \mathrm{T}_{\mathrm{em}} - \mathrm{T}_{\mathrm{damp}}\right) \tag{1}$$

Replacing T_{em} and performing a small perturbation on all variables about an operating point where the rotor angle is at δ_0 . We obtain,

$$\frac{1}{\omega_{b}}\frac{d^{2}\delta}{dt^{2}} = \frac{1}{2H} \{ (\Delta T_{mech} + \frac{E'V\omega}{X}\cos\delta_{0}\Delta\delta - D_{\omega}\Delta(\frac{(\omega_{r}-\omega_{e})}{\omega_{b}}) \}$$
$$K_{S} = \frac{E'V\omega}{X}\cos\delta_{0}\Delta\delta$$
(2)

where K_s is termed the synchronizing torque coefficient. In the Laplace domain, the above differential equation becomes

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$$s^{2}\Delta\delta + \frac{D_{\omega}\omega_{b}}{2H}s\Delta\delta + \frac{K_{s}\omega_{b}}{2H}\Delta\delta = \frac{\omega_{b}}{2H}\Delta T_{mech}$$
(3)

Comparing the characteristic equation portion of equation (3) with that of normal second order equation, $s^2 + 2\zeta\omega_n s + \omega_n^2$, with roots $(-\zeta \pm j\sqrt{\zeta^2 - 1})\omega_n$, the expressions for the undamped or natural frequency, ω_n and the damping coefficient, ζ are

$$\omega_n = \sqrt{\frac{K_s \omega_b}{2H}}$$

$$\zeta = \frac{\omega_b}{2\omega_n} \left(\frac{D_\omega}{2H}\right) = \frac{1}{2} \frac{D_\omega}{\sqrt{2H\omega_b K_s}}$$
(4)

Note that K_s is inversely proportional to X, the total reactance between E' and V_{∞} ; in other words, the synchronizing torque coefficient will increase with electrical strength of the network at generator bus. According to the equations (4), a higher K_s raises the value of ω_n , but lowers ζ , whereas a higher H will decrease the value of both ω_n and ζ .

The linear model of the system about the given operating point is developed, and from the linear model, the dominant eigen values of the system can be determined. Following which the simulation runs are conducted. The linear or state variable model of a multi-machine system (Two generator infinite bus system) is as below:

$$\dot{x} = Ax + Bu$$

$$y = Cx$$
 (5)

The system matrix A is function of the system parameters, which depends on the operating conditions. The perturbation matrix B depends on the system parameters only. The perturbation signal u is ΔT_{mech} . The output matrix C relates the desired output signals vector y to the state variables vector x.

B. Modeling of Power System Stabilizer

Fig 2 shows the block diagram of the excitation system including AVR and PSS. It consists of three blocks: a phase compensation block, a signal washout block, and a gain block. The phase compensation block provides the appropriate phase-lead characteristic to compensate for the phase lag between the exciter input and the generator electrical (air-gap) torque.



Fig. 2. Block diagram representation of conventional PSS

The signal washout block serves as a high pass filter, with the time constant T_w high enough to allow signals associated with oscillations in ω_r to pass unchanged. Without it, steady changes in speed would modify the terminal voltage. It allows the PSS to respond only to changes in speed. From the viewpoint of the washout function, the value of T_w is not

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critical and may be in the range of 1 to 20 seconds [11]. The main consideration is that it is long enough to pass stabilizing signals at the frequencies of interest unchanged, but not so long that it leads to undesirable generator voltage excursions during system islanding conditions. The stabilizer gain K_s determines the amount of damping introduced by the PSS. Ideally, the gain should be set at a value corresponding to maximum damping. Since the purpose of a PSS is to introduce a damping torque component, a logical signal is to use for controlling generator excitation is the speed deviation $\Delta \omega_r$.

III. LINEAR QUADRATIC REGULATOR

The LQR problem, where all the states are known is the deterministic initial value problem: Given the system $\dot{x} = Ax + Bu$ with a given non-zero initial state x(0), find the input signal u(t) which the system back to the zero state (x=0) in an optimal manner, i.e., by minimizing the deterministic cost

$$\mathbf{J}_{\mathbf{r}} = \int_0^\infty \left(x(t)^T Q x(t) + u(t)^T R u(t) \right) dt \tag{6}$$

The optimal solution is for any initial state $u(t) = -K_r X(t)$ where $K_r = R^{-1}B^T X$ and $X = X^T >= 0$ is the unique positive semi definite solution of the algebraic Riccati equation, $A^T X + XA - XBR^{-1}B^T X + Q = 0$ (7)

C. Kalman filter

The Kalman filter has the structure of an ordinary state estimator or observer with

$$\dot{\hat{\mathbf{x}}} = \mathbf{A}\hat{\mathbf{x}} + \mathbf{B}\mathbf{u} + \mathbf{K}_{\mathbf{f}}(\mathbf{y} - \mathbf{C}\hat{\mathbf{x}}) \tag{8}$$

The optimal choice of K_f , which minimizes $E = \{[x - \hat{x}]^T [x - \hat{x}]\}$ is given by,

$$\mathbf{K}_{\mathrm{f}} = \mathbf{Y}\mathbf{C}^{\mathrm{T}}\mathbf{V}^{-1} \tag{9}$$

Where $Y = Y^{T} \ge 0$ is the unique positive semi definite solution of the algebraic Riccati equation,

 $YA^{T} + AY - YC^{T}V^{-1}CY + W = 0$ (10)



Fig. 3. The separation theorem

D. LQR Procedure

- 1) Choose Q and R such that $Q = C^{T}C$, with (A; C) detectable, and $R = R^{T} > 0$
- 2) Solve the Riccati equation, $PA + A^{T}P + Q - PBR^{-1}B^{T}P = 0$ (11)
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3) Simulate the initial response of $\dot{X} = (A + BK)X$ for

and compute $\mathbf{K} = -\mathbf{R}^{-1}\mathbf{B}^{\mathrm{T}}\mathbf{P}$

different initial conditions4) If the transient response specifications and/or the magnitude constraints are not met, go back to step 1 and re-choose O and/or R.

The optimal PSS is obtained using LQR technique for the plant. Eigen value based stability analysis is made for the system with and without LQR based PSS is listed in the Table 2. It is observed that the system with LQR based PSS has less oscillatory modes and the Eigen values of the system are shifted to the left hand side of PSS when compared to that of the system without PSS.

IV. $H\infty$ LOOP SHAPING DESIGN

 H_{∞} loop shaping design procedure follows three steps,

- Loop shaping
- Robust Stabilization



Fig. 4. Hinf loop shaping procedure

E. Application of Optimization Technique

The Robust PSS is developed using $H\infty$ loop shaping design procedure. The selections of weights are made using the optimization process. The weights are selected using the formula,

$$w_{1} = (s/M + w_{b}) / (s + (w_{b}*a));$$

$$w_{2} = ((L*s) + 1) / (2*(0.5*L*s+1));$$
(12)

such that, the steady sate offset is less than 'a'; closed loop bandwidth is higher than w_b ; Amplification of high frequency noise is less than a factor M, are chosen as decision variables.

The weights are tuned using GA. With objective function as the minimization of infinite norm of sensitivity and complementary sensitivity function,

$$F = \min \left\| \frac{S}{T} \right\|_{\infty}$$

$$S = (I+G^*K)^{-1};$$

$$T = G^*K^*(I+G^*K)^{-1};$$
(13)

where G is the nominal plant and K is the optimal controller obtained using normalized co-prime factorization method. Table 1 shows the GA parameters for the optimization process.

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TABLE I Ga Parameters					
1	Number of decision variables	4			
2	Population size	50			
3	Iteration	50			
4	Selection rate	0.5			
4	Crossover probability	0.6-1.0			
5	Mutation rate	0.05			

The convergence characteristics of GA for getting optimal weights are as shown in Fig 5.



Fig. 5. Convergence characteristics of GA

The Optimal weights chosen from the optimization algorithm are as follows,

$$W_1 = 0.7*(s+6076)/(s+0.2)$$
$$W_2 = 1*(s+83)/(s+160)$$
(14)

The singular value graph of the nominal plant, desired plant, and weights W_1 and W_2 are as shown in Fig.6 Stronger Values



Fig. 6. The singular value plots of the G, w1, w2, and Gd



Fig. 7. The singular value plots of G*K, G and margins of Gd (Gd/GAM and Gd*GAM)

It is observed from the graph that the weights are finely tuned in such a way the closed loop response of the plant G^*K is within the margin of the desired plant.

The gain of the desired plant is increased in order to damp the low frequency oscillations. K_{∞} controller is designed to achieve robust stability for the plant. The maximum stability margin is $\epsilon_{max}=0.2741$. This margin evaluates the feasibility of our loop shaping design. The normalized left co-prime factorization of the nominal plant as $G_{so} = \widetilde{M_s}^{-1}\widetilde{N_s}$, the controller K_{∞} can stabilize all $G_s = (\widetilde{M} + \Delta_M)^{-1}(\widetilde{N} + \Delta_N)$ satisfying $||\Delta_M, \Delta_N||_{\infty} < 0.2741$. The final controller is the combination of W_1 and W_2 with K_{∞} that is K=W₁* K_∞*W₂.

V. SIMULATION RESULTS

The Simulink model for Two Machine Infinite bus system is developed. The model for Machine1 with PSS is shown in Fig.7. The mechanical and electrical equations of generator are modeled inside the rotor and stator_wdg block. In this model, the transformations of the input stator abc voltages to the rotor qd reference frame are performed inside the qde to qdr block. The transformation uses the $\cos\theta_r(t)$ and $\sin\theta_r(t)$ generated by the oscillator block. The VIPQ block is generated in which the instantaneous magnitude of the stator voltage, stator current, and stator real and reactive power at the generator's terminal are computed. The instantaneous values of the real and reactive power in per unit flowing out of the stator terminals of the generator are computed from the following equations:

$$P = V_d i_d + V_q i_q$$

$$Q = V_q i_d - V_d i_q$$
(15)

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Fig. 8. Simulink model for Machine1 with PSS

The stability condition of the system can be examined using their eigen values. Hence the system eigen values are determined and listed in Table 2. In Table 2, first column shows the eigen values of the system with the presence of conventional Power System Stabilizer and second column shows the eigen values for the system with the presence of PSS modeled based on LQR technique. From the table, it proves that the oscillations are reduced in the system using LQR based PSS since four oscillatory eigen values are damped out completely and also some of the eigen values moves to the large negative value. It shows that the system moves towards the stability region quickly while using LQR technique.

TABLE II

		Conventional Pss	Based Pss	
	1	-8.4390±35.5645i	-8.6114±35.7086i	
	2	-2.1336±12.2426i	-7.0035±13.7044i	
	3	-8.5958± 8.2069i	-8.4708±8.3773i	
	4	-0.5325±7.5256i	-6.3476±9.5902i	
	5	-0.5084±0.7588i	-6.9828	
	6	-0.2481±0.2575i	-5.7374	
	7	-6.6920	-2.9622	
	8	-5.5063	-1.9390	
	9		-1.3507	
	10		-1.0509	
slip1(p.u)	3 x 10 ⁻³ 2 1 1 -1 -2 -3	with LQR based PSS	1.4	PSS
	-4	10 15 20 25 30	0.4 5 10 15 20 25	3(

EIGEN VALUES OF THE SYSTEM WITH CONVENTIONAL AND LQR BASED PSS

Eigen Values Of The

System With Lqr

time (s)

Eigen Values Of The

System With

S.No.

Fig. 9. Response of rotor slip and real power of machine1

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time (s)



Fig. 10. Response of Real power and rotor slip of machine2



Fig. 11. The response of rotor slip (i)Without PSS and Robust PSS and (ii)LQR based PSS and Robust PSS

Simulation is done for three different cases: (i) System without PSS, (ii) System with LQR based PSS and case (iii) System with Robust PSS. All parameters of the system are analyzed after the simulation is carried out.

Among that, the response of rotor slip and real power of both machines are shown in Fig.8 and Fig.9. In these Figures, comparison is made between two cases: system without PSS and system with LQR based PSS. The analysis is made with the T_{mech} disturbances i.e., the magnitude of T_{mech} is changed from 0.8 p.u to 0.9 p.u.at time t=15s. So at this instant, there are some oscillations present in the system. But with the use of LQR based PSS, the oscillations are damped out quickly which proves in the fig.8 and Fig.9. Hence it is observed that the LQR based PSS damps out the oscillation well even under perturbations.

Comparison is also made between LQR based PSS and Robust PSS which shown in Fig.10. The response of the rotor slip of machine1 without PSS and with Robust PSS is as shown in first Fig.10. It is observed from the above graph that the response of rotor slip in machine1 with Robust PSS has faster response compared to that of the system without PSS. In case of comparison with the LQR based PSS it is observed that there is only slight difference in damping out oscillations. All the remaining electrical parameters such as Voltage, Current, Reactive Power, Torque and angular speed are also analyzed for all the three cases.

VI.CONCLUSION

A systematic approach to design PSS using LQR technique and H_{inf} loop shaping procedure is presented for two generator infinite bus system. The analysis is made to verify the robustness of the designed controller using loop shaping procedure. Genetic Algorithm is applied to determine the weights while designing robust controller. The resulting PSS can stabilize the nominal plant. Proposed work demonstrates the good damping performance of the designed controller and the comparison is made between LQR based PSS and Robust PSS. The LQR based PSS is tested under perturbations such as change in mechanical torque and found to have better damping performance even under perturbation.

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The above procedure can be applied to large multi-machine to design the robust controller to take care of the oscillations under perturbed conditions.

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