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# Circularly Polarized Slit Loaded Broadband Coplanar Capacitive Coupled Probe Fed Microstrip Antenna for Wireless Communication System

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**ABSTRACT:** In this paper we propose a circularly polarized coplanar capacitive fed rectangular patch Microstrip antenna loaded with I-shape slots and two slits within the patch. The model of Antenna uses the capacitive feed strip which is fed by coaxial probe using equivalent circuit approach. The circular polarization is achieved by introducing two slits of equal dimension to produce the orthogonal modes. The proposed antenna has an axial ratio bandwidth (<3dB) of 5.08% and impedance bandwidth (S<sub>11</sub><10dB) of 60% in the frequency range 3.82GHz to 7.11GHz which shows wideband characteristics. Gain is almost constant in the entire band of operation. The proposed antenna is simulated using IE3D software. It is shown that the antenna can cover the bands of several applications including WLAN, Satellite communication, Marine communication and terrestrial microwave links.

KEYWORDS: Microstrip Antenna, capacitive coupling, circular polarization, slot, capacitive coupling.

#### I. INTRODUCTION

A Microstrip antenna is attractive candidate in the variety of commercial scale application such as mobile and spacecraft application due to their numerous advantages such as low cost, light weight and easily printed onto circuit board. However Microstrip antennas suffer from low bandwidth characteristics [1].Many efforts have been devoted to bandwidth widening techniques of Microstrip antennas, including the use of impedance matching [2], multiple resonators [3], and a thick substrate [4]. The simplest technique to overcome the size problem is to incorporate high dielectric constant substrate which is very expensive so it is not suitable in many applications [5]. The other technique using short circuit [6] or a short pin [7] has also been proposed. Much attention is focused to improve the bandwidth of short-circuited patch antennas [8-10].18.5% impedance bandwidth was obtained using thick foam as substrate [10].Two stacked short circuited patch were used to obtain impedance bandwidth of 30% [9].Using an L-probe feed substrate the impedance bandwidth obtained was more than 39% [10].Thus stacked and L-probe feed both increase the complexity of the antenna. In case of using high dielectric substrate the inductance introduced by coaxial probe will increase the height of foam which in turn limits the bandwidth improvement of the antenna. The U-slot cut on the patch proposed [11] can be used to cancel the inductance.

A single layer coplanar capacitive fed wideband Microstrip antenna is easy to fabricate and reported 28% impedance bandwidth [12] could be enhance by optimizing the feed strip dimensions and its position with respect to radiator patch [13]. The probe inductance is compensated by placing capacitive feed strip along one side of radiator patch. The input impedance of Microstrip antenna is a key factor for matching between transmission line and antenna terminal. For impedance calculation various models like transmission line model, cavity model, moment method solution and equivalent circuit are used in which the equivalent circuit model [14], is simple and reasonably accurate but they consider that the probe feed is directly connected to radiator patch.

Circularly polarized antennas are required for radar tracking GPS systems and satellite communication systems because circular polarization provides reduction in multipath effect and they are relatively insensitive to the transmitter and Copyright to IJIRSET www.ijirset.com 12917



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receiver orientation [15] than for linearly polarized antennas. Although Microstrip antennas in their basic form normally provide linear polarization, circular polarization can be obtained by certain modification to the basic antenna geometry and /or feed. These modifications includes adjusting the dimensions of basic patch with one or more feed, trimming the corners of the square patch, feeding the patch at adjacent sides, cutting the slots inside the patch, and feeding the patch(rectangular) from its corners along the diagonals[16].Several CP antennas available in this literature. In conventional single feed circularly polarized Microstrip antenna design, a substrate with high permittivity s often used to make the antenna compact[17] but it has an inherent drawback of low axial ratio bandwidth. The antenna reported in [18] has axial ratio band and impedance matching band does not overlap when the substrate thickness is increased [19]. The antenna reported by [19] has axial ratio bandwidth of 6.3% with straight feed and increases up to 14.1% with L-strip feed which is a corner trimmed CP antenna with U-shaped slot cut on the patch to insure wide impedance bandwidth. The antenna reported in [20] and [21] are having axial ratio bandwidth of 14% and 23% respectively.Since these antennas consist of stacked configurations and hence they are very difficult to fabricate. The probe feed H –shape Microstrip antenna fed along with its diagonal to get circular polarization[22] have axial ratio bandwidth of just 1.3%.

Here we have proposed a design of I-shape slot loaded capacitive coupled patch antenna that can have wideband operation. The slot increases the impedance bandwidth. The I-shape slot introduces a capacitance suppressing some of inductance introduced by probe. Two slits of equal dimension introduced in the rectangular patch are responsible for exhibiting circular polarization. The characteristics including impedance bandwidth, axial ratio bandwidth surface current distribution, gain, radiation pattern has been discussed. The proposed antenna has impedance bandwidth of 60% in the frequency range 3.82GHz to 7.11GHz which shows the wideband behaviour. The antenna has axial ratio bandwidth of 5.04% in frequency range 6.69GHz to 7.05GHz The antenna is useful for wideband applications including, RADAR, Marine Satellite communication, WLAN and terrestrial microwave links..

### **II. ANTENNA GEOMETRY AND DESIGN**

The basic geometry of the antenna is shown in fig.1. The proposed antenna is basically a suspended Microstrip antenna in the air with a radiating patch and the feed strip which are placed above the substrate of thickness 'h' mm. The length of radiator patch (L) is 15.5mm and the width of patch (W) is 16.4mm. Capacitive coupling of energy to the radiating patch is performed by feed strip using a long pin SMA connector The capacitive coupling is given by feed strip with length of strip (s) is 3.7mm and width of strip (t) is 1.2mm. The feed strip is separated from patch by distance (d) is 0.5mm. The substrate used for the antenna fabrication is Fr4 of dielectric constant 4.4 and of thickness (h) of 1.56 mm. The air gap between substrates is taken 6mm. Thus the effective height of the substrate is the total height of the air gap and the height of effective dielectric constant s. The effective dielectric properties will remain almost unaffected by small variation of the dielectric constant of the substrate and so the bandwidth also. The slot is embedded on the patch with length of 12 mm and width of 2mm at 4mm above from centre of the rectangular patch. By introducing the slot the impedance bandwidth increases. The variation in slot length affects the impedance bandwidth and also shifts the corner frequency. Two slits of length (L<sub>slit</sub>) of 4mm and width (W<sub>slit</sub>) of 1mm has been introduced at 2mm above centre of patch which results in circular polarization.





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Fig.1.Basic geometry of slot loaded capacitive feed wide band Microstrip antenna (a)Top view (b) Cross sectional view

This Microstrip antenna is equivalent to a parallel tuned circuit for antenna operating in the fundamental mode and shown in fig.2 which consists of parallel RLC circuit [13].



Fig.2. Equivalent circuit of probe with feed strip

In present geometry of antenna the feed is not given by directly connecting the coaxial probe, but the radiator patch is connected to a small rectangular patch which excites it by capacitive coupling. The feed strip is equivalent to rectangular Microstrip capacitor because the dimension of the feed strip is much smaller in comparison to wavelength of operation. Since the area of the strip is very small as compared to patch, the perimeter to area ratio is much larger hence the fringing capacitance from the edges may be significant in this case. This capacitance is known as terminal capacitance of the line.

The dissipation and radiation losses from feed strip line can be represented as  $R_{strip}$ . At lower frequency radiation losses from feed strip line are negligible but as the operating frequency increases, the radiation losses are effective. The separation between radiator patch and the feed strip is essentially an asymmetrical gap that means two conductors of unequal width separated by short distance (d). This can be represented by model of Cp1, Cs, and Cp2. Here Cp1 and Cp2 represent terminal capacitance of two Microstrip sections and the series capacitance Cs represents gap [13]. The slit introduced in the patch, the resonance frequency will change due to change electrical size of antenna. This effect is represented by a parallel combination of inductance  $L_{slit}$  and  $C_{slit}$ [23] and equivalent circuit of feed strip is shown in figure 2.



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Fig.3. Circuit model of Antenna without slot

The expression of slit inductance is given by

where

$$L_{s} = \frac{60h}{c} \ln\{\frac{c}{0.2886\pi f_{r}}\}$$
(1)

Where c is the velocity of light and  $f_r$  is the resonance frequency. The capacitance offered by the slits can be calculated as gap capacitance by considering the two wings of the Microstrip antenna as coupled Microstrip lines of length  $l_p[23]$ 

$$C_{s=2} l_{n} \frac{\epsilon_{0}}{\pi} \left[ \ln \left( 2 \frac{1 + \sqrt{k'}}{1 - \sqrt{k'}} \right) + \ln \coth \left( \frac{\pi s}{4h} \right) + 0.013 C_{f} \frac{h}{s} \right] F \quad (2)$$

$$k'_{=} \sqrt{1 - k^{2}}$$

$$k = \frac{1 + \frac{w_{1}}{s} + \frac{w_{2}}{s}}{(1 + \frac{w_{1}}{s})(1 + \frac{w_{2}}{s})}$$

The input impedance of antenna without slot as shown in fig.3 can be calculated from equation (1)

$$Z_{\text{in}} = \left[\frac{1}{Z_{patch}} + j\omega C_{p1}\right]^{-1} + \frac{1}{j\omega C_{ps}} + \left[\frac{1}{Z_{feed\_strip}} + j\omega C_{p2}\right]^{-1} + jX_{probe} + Z_{\text{slit}} (3)$$

#### **III. RESULTS AND DISCUSSIONS**

The results of the proposed antenna were obtained by simulation using the IE3D software which is a method of moments (MoM) based electromagnetic software. Figure 5 shows variation of return loss with frequency. Fig.5 shows that -10 dB return loss is between the frequency range 3.82GHz to 7.11GHz. The center frequency of this proposed design is 5.46GHz and the impedance bandwidth is 60% which shows that the antenna exhibits broadband characteristics.

An attempt has been made to study the behaviour of antenna on slot dimensions. Variation of return loss with frequency for different slot widths is given in figure 5(b). The slot width have very small effect on impedance bandwidth and axial ratio bandwidth and it shifts the centre frequency very slightly. The variation of return loss with frequency for different slot lengths is shown in figure 5(c) which shows that by increasing the slot length impedance bandwidth and center frequency both decreases slightly.

Axial ratio variation with frequency has been shown in fig. 6 in which the axial ratio is less than 3dB in the frequency range from 6.69GHz to & 7.05GHz .Thus the axial ratio bandwidth is 5.05% which shows the property of circular polarization.



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Fig. 5(a) Variation of return loss with frequency (Simulated)



Fig 5(b): Variation of return loss with frequency for different slot widths.(Simulated)



Fig.5(c) Variation of return loss for different slot lengths (Simulated)

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Fig.6. Variation of axial ratio with frequency (Simulated)

The surface current distribution at 3.80GHz, 5.44GHz and 7.12 GHz has been shown in figure 7(a), 7(b) and 7(c) respectively. The figure shows that electric current moves around the slot hence increase the path of the electric current and responsible for the impedance bandwidth improvement.

The gain variation with frequency is shown in figure 8. The variation of gain is less than 1dB. The radiation patterns at frequencies 3.84GHz, 5.44GHz and 7.12GHz are shown in figure 9(a), 9(b), 9(c). The patterns shows that H plane patterns are symmetrical throughout the band of operation whereas E plane patterns are symmetrical at lower frequencies and as the frequency increases, the degree of asymmetry increases.



Fig. 7(a).Surface current distribution at frequency 3.8GHz



Fig. 7(b) Surface current distribution at frequency 5.44GHz



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Fig.7(c) Surface current distribution at frequency 7.12GHz



Fig.8.Variation of gain with frequency (Simulated)











(An ISO 3297: 2007 Certified Organization) Vol. 3, Issue 5, May 2014 f=7.08(GHz) E-Plane f=7.08(GHz), H-Plane 0.0 ae 9.0 ్లం (180-**q**) -2.0 -6.0 -10.0 -14.0 -18.0 -14.0 -10.0 -2.0 2.0 -6.0 2.0 130.0 0'091 0.091 0.081 Elevation Pattern

Fig.9(c) Radiation Pattern at frequency 7.12GHz

#### **IV.CONCLUSION**

In this communication, circularly polarized wideband slot loaded capacitive coupled fed Microstrip antenna with broad band behaviorhas been presented. The antenna has achieved 5.04% axial ratio bandwidth and impedance bandwidth of 60% in the frequency range from 3.82GHz to 7.11GHz. This antenna can be used for wireless communication system. The proposed antenna having simple structure and operates in 5GHz frequency band hence can be used for WLAN application. The antenna can cover the bands of several applications including WLAN, Satellite communication, Marine communication and terrestrial microwave links.

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