



Comparison and Analysis of Outage Probability & Signal Power with Interferences at Different Fading Environment

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Abstract: In this paper a mathematical method is presented to express the outage probability for a desired radio signal received from a mobile transmitter in the presence of single or multiple interfering signals with Rician and Rayleigh fading. In contrast to previously reported analysis, this paper compares the outage probability & signal power at different fading environment. This is useful in determining the spectrum efficiency and performance of (interference-limited) radio networks for cellular telephony, two-way paging, and another mobile data networks. For this analysis the interference may be single or multiple. In this paper we focus on the different parameters for output value of the different signals.

Keywords – Outage probability, Rician Channel, Rayleigh Channel, Interferences.

I. INTRODUCTION

In mobile communication, the quality of service is often expressed in terms of the outage probability experienced by subscribers near the boundary of the service area of a base station. Due to limited spectrum availability, radio networks become more and more limited by mutual interference between subscribers. Received signal at the receiver is affected due to reception of no. of multipath signals. These signals are Rician-faded & the co-channel interferers are modeled to be Rayleigh-faded. This paper is based on the consideration that a direct line-of-sight signal component exists in within-cell transmission, while the interferers are assumed to be Rayleigh-faded because a direct LOS path between co-channel cells is improbable to exist. Closed-form outage probability statements are derived for both single and multiple interferences.

The performance analysis of mobile systems is reported in numerous research papers. Since microcells provide enhanced spectrum efficiency and economical power of handhelds, recently researchers have turned their attention to work over outage probability and spectrum efficiency of microcellular radio. In the outage probability is evaluated, assuming the desired signal as Rician fading, and considering interfering signals as Rayleigh distributed without in view of the path-loss law. The spectrum efficiency is calculated in [7], considering both desired and interfering signals as Rician faded.

II. SYSTEM MODEL

We assume the limiting source of performance degradation [1]; therefore, for simplicity, we reject the thermal noise in our system model and consider only an interference limited environment. The transmitted signals from the desired and the interfering subscriber are,

$$S_s(t) = \sqrt{P_s} T \sum_{M=-\infty}^{M=\infty} a_s[m] h_\tau(t - mT)$$

And

$$S_i(t) = \sqrt{P_i} T \sum_{M=-\infty}^{M=\infty} a_i[m] h_\tau(t - mT)$$

where h_τ is transmitter pulse response, $1/T$ is the data transmission rate, and P_s and P_i are the transmitting powers of the desired and the i^{th} interfering signals, respectively

III. OUTAGE PROBABILITY FOR RICIAN FADING ATMOSPHERE

Outage probability is a key figure of merit in wireless communications. In interference-limited systems, outage probability is commonly defined as the probability that the signal-to-interference ratio (SIR) of a received signal is below a given threshold. In this part, we obtain a closed-form expression for this case.



Derivation of Outage Probability for Rician Fading Channels

Let X_k ($k = 0, \dots, L$) be a set of independent complex Gaussian random variables, with mean value m_k and variance σ_0^2 . Let X_0 represent the desired Rician signal and $X_k, k=1$ represent the sum of L Rician interferers. The outage probability for L interferers is expressed as

$$P(\text{outage } L) = \Pr(Y \leq 0) \dots\dots\dots(1)$$

Where

$$Y = Y_0 - \sum_{k=1}^L \dots\dots\dots (2)$$

And $Y_0 = |X_0|^2$ and $Y_k = R_l |X_k|^2$; $k = 1, \dots, L$.

Here, R_l is the signal to interference protection ratio. To derive $P(\text{outage}|L)$ given in (1), it is necessary to find the probability density function of Y . For this, it is convenient to find the characteristic function of Y . Turin has evaluated the characteristic function of a set of N complex Gaussian random variables. In our case, we need to adopt Turin's result, but only for each of the ($k = 0, \dots, L$), we have

$$F_k(s) = \frac{\exp\left(\frac{s C_k}{1 - s \beta_k}\right)}{1 - s \beta_k} \dots\dots\dots(3)$$

Where $C_0 = |m_0|^2$, $\beta_0 = \sigma_0^2$, $C_k = R_l |m_k|^2$, $\beta_k = R_l \sigma_0^2$, $k = 1, \dots, L$. It is noted that C_0 and β_0 respectively, the average power of the specular and diffused components of the desired Rician signal, whereas C_k and β_k are, respectively, the average power multiplied by the protection ratio of the specular and diffused components of the interferer. Now, assuming that all the signals X_k are independent the characteristic function, $F(s)$ of Y is given by

$$F(s) = F_0(s) \prod_{k=1}^L F_k(-s) \dots\dots\dots(4)$$

If we assume further that all the interferers have the same mean $m_k = m_l$ and the same variance $\sigma_k^2 = \sigma_l^2$, which also implies that the Rice factor of each of the interfering signals is the same, $K_l = |m_l|^2 / \sigma_l^2$, then

$$F(s) = F_0(s) [F_k(-s)]^L \dots\dots\dots (5)$$

The pdf of the sum of the interferers $Y_l = \sum_{k=1}^L Y_k$ where Y_k is defined in (2), can be obtained as the inverse Fourier transform of $[F_k(-s)]^L$. The inverse Fourier transform of was obtained by Bello, who applied the result of Campbell and Foster. The result is shown below:

$$P_{Y_l}(Y_l) = \frac{1}{2\pi j} \int [F_k(-s)]^L e^{-s Y_l} ds$$

$$= \begin{cases} \frac{1}{\beta_k} \left(\frac{Y_l}{L C_k}\right)^{L-1/2} \exp\left(-\frac{Y_l + L C_k}{\beta_k}\right) \\ \times I_{L-1}\left(\frac{2}{\beta_k} \sqrt{Y_l L C_k}\right), & Y_l > 0 \\ 0 & Y_l < 0 \end{cases} \dots\dots\dots (6)$$

The pdf of the desired signal is a special case of (6), with $L = 1$ and $K = 0$:

$$P_{Y_0}(Y_0) = \begin{cases} \frac{1}{\beta} \exp\left(-\frac{Y_0 + C_0}{\beta_0}\right) I_0\left(\frac{2}{\beta_0} \sqrt{Y_0 C_0}\right), & Y_0 > 0 \\ 0, & Y_0 < 0 \end{cases} \dots\dots\dots (7)$$

Where (x) is the modified Bessel function of the first kind and order m . The pdf of Y and $p(Y)$, defined in (2), can be found by convolving (6) and (7). The outage probability is then

$$P(\text{outage } L) = \int_{-\infty}^0 p(Y) dY$$

$$= \int_{-\infty}^0 dY \int_0^{\infty} p_{Y_1}(Z) p_{Y_0}(Z+Y) dZ, \dots (8)$$

Now, making the change of variables, $Y = \gamma - Z$, in (8) and interchanging the order of integration, we have

$$P(\text{outage } L) = \int_0^{\infty} \frac{1}{\beta_K} \left(\frac{Z}{LC_K} \right)^{\frac{L-1}{2}} \left\{ \begin{array}{l} x \exp\left(-\frac{Z+LC_K}{\beta_K}\right) I_{L-1}\left(\frac{2}{\beta_K} \sqrt{Y_1} LC_K\right) \\ x \int_0^Z \frac{1}{\beta_0} \exp\left(-\frac{\gamma+C_0}{\beta_K}\right) \\ x I_0\left(\frac{2}{\beta_0} \sqrt{\gamma C_0}\right) d\gamma dZ \quad \dots \dots \dots (9) \end{array} \right.$$

Again, making the change of variables

$$Z = \frac{\beta_K x^2}{2}; \quad \gamma = \frac{\beta_0 y^2}{2} \dots \dots (10)$$

In (9) by defining the parameters

$$LC_K = \frac{\beta_K a^2}{2}; \quad C_0 = \frac{\beta_0 b^2}{2}; \quad r = \sqrt{\frac{\beta_K}{\beta_0}} \dots \dots (11)$$

Where $a^2/2 = K_I L$ and $b^2/2 = K_0 =$ Rice factor of the desired signal, we arrive at

$$P(\text{outage } L) = a^{-L+1} \int_0^{\infty} x^L \exp\left(-\frac{x^2+a^2}{2}\right) I_{L-1}(ax) \\ \times \int_0^{rx} y \exp\left(-\frac{y^2+b^2}{2}\right) I_0(by) dy dx \dots (12)$$

The double integral is a special case of the integral evaluated by Price [9, eq. (2.5)]. Using [9, eq. (3.23)] and [9, eq. (3.24)], the new closed-form expression of the outage probability for the Rician fading channel is found to be

$$P(\text{outage } L) = Q\left[\sqrt{\frac{2LK_1R_1}{b_1+R_1}}; \sqrt{\frac{2K_0b_1}{b_1+R_1}}\right] + \\ \exp\left(-\frac{LK_1R_1+K_0b_1}{b_1+R_1}\right) \\ \times \sum_{m=0}^{L-1} \left(\frac{K_0R_1}{LK_1R_1}\right)^{m/2} I_m\left(\sqrt{\frac{4LK_1K_0b_1R_1}{b_1+R_1}}\right) \\ \times \left\{ \left(1 + \frac{b_1}{R_1}\right)^{-L} \sum_{k=m}^{L-1} \binom{L}{k-m} \left(\frac{b_1}{R_1}\right)^k - \delta_{m0} \right\} \\ m \geq 0 \dots \dots \dots (13)$$

where $b_1 = \sigma_0^2/\sigma_I^2$ and $\delta_{m0} = 1$ for $m=0$, $\delta_{m0} = 0$ for $m \neq 0$, and the Marcum's Q function is

$$Q(u,v) = \int_v^{\infty} x \exp\left[-\frac{1}{2}(x^2+u^2)\right] I_0(xu) dx \dots \dots (14).$$

3. DERIVATIONAL RESULTS FOR DIFFERENT CASES

1) In first case if we assume that Desired Signal is the Rayleigh Signal and Single Rician Interferer, and $L = 1$ and $K_0 = 0$ then according to equation (13)

$$P(\text{outage } L) = 1 - \frac{b_1}{R_1 + b_1} \exp \left[- \left(\frac{K_1 R_1}{R_1 + b_1} \right) \right] \dots (15)$$

2) In second case if we assume that Desired Signal is the Rician Signal and Single Rayleigh Interferer, and $K_1 = 0$ and $L = 1$, then according to equation (13)

$$P(\text{outage } L) = \frac{R_1}{R_1 + b_1} \exp \left[- \left(\frac{K_0 b_1}{R_1 + b_1} \right) \right] \dots (16)$$

3) In third case if we assume that Desired Signal is the Rician Signal and Single Rician Interferer and $L = 1$, then according to equation (13)

P (outage L)

$$= Q \left[\sqrt{\frac{2 K_1 R_1}{b_1 + R_1}} ; \sqrt{\frac{2 K_0 b_1}{b_1 + R_1}} \right] - \frac{b_1}{b_1 + R_1} \exp \left[- \frac{K_1 R_1 + K_0 b_1}{b_1 + R_1} \right] \times I_0 \left(\sqrt{\frac{4 K_1 K_0 b_1 R_1}{b_1 + R_1}} \right) \dots (17)$$

In equation (13), we have assumed that all the interferers have the same mean $m_k = m_l$ and $k = 1, \dots, L$. If all the interferers have distinct means m_k , but with identical variance $\sigma_k^2 = \sigma_l^2$, the Rice factors of the L interferers are distinct with $K_k = m_k^2 / \sigma_l^2$ and $k = 1, \dots, L$.

In equation (6) and (9), we now have the parameter $\sum_{k=1}^L C_k$ instead of LC_k . Note that this holds, even when some of the interferers are Rayleigh-faded, i.e., $K_k = 0$, for certain k. Consequently, the outage probability for the case, where all the interferers have distinct means with the parameter LC_k now being substituted by $\sum_{k=1}^L K_k$. We can see that the outage probability for this case is dependent on the sum of LOS power components of the L Rician interferers.

IV. COMPUTATIONAL RESULTS FOR DIFFERENT CASES

Figs. 1 and 2 show the outage probability graphs as a function of the signal-to-interference power ratio (SIR), and defined as

$$SIR = \frac{(K_0 + 1) \sigma_0^2}{L(K_1 + 1) \sigma_l^2} = \frac{(K_0 + 1)}{L(K_1 + 1)} b_1$$

It is to be noted that when comparing the curves of single and multiple interferers, for a given SIR, and given K_0 , σ_0^2 , and K_1 and the value of $L\sigma_l^2$ is then fixed. Therefore, for the larger L value, the corresponding σ_l^2 is smaller. This means that the power of the single interferer is distributed among the L interferers. In fig 1, Outage probability curves for $R_l = 5$ and for various values of K_0 , K_1 and L are shown.

By comparing the curves for $K_0 = K_1 = 10$, with $L = 1$ and $L = 6$, it can be seen that when the power of a single interferer is distributed among L interferers, a lower P (outage/L) can be expected. At the other extreme, when the desired signal is Rayleigh-faded and the interferer is Rician faded ($K_0 = 0$ and $K_1 = 5$) the P (outage/L) is much larger than the other previous curves.

Fig. 2 shows the effect of different R_l and L on the outage probability for Rayleigh signal. According to graph the outage probability increases as R_l increases. It can also be seen that a larger L gives a smaller outage probability, as observed in Fig. 1.

Fig. 3 shows the effect of different R_l and L on the outage probability at $K_l = 0$ for Rician signal. According to graph the outage probability is similar to the Fig. 2.

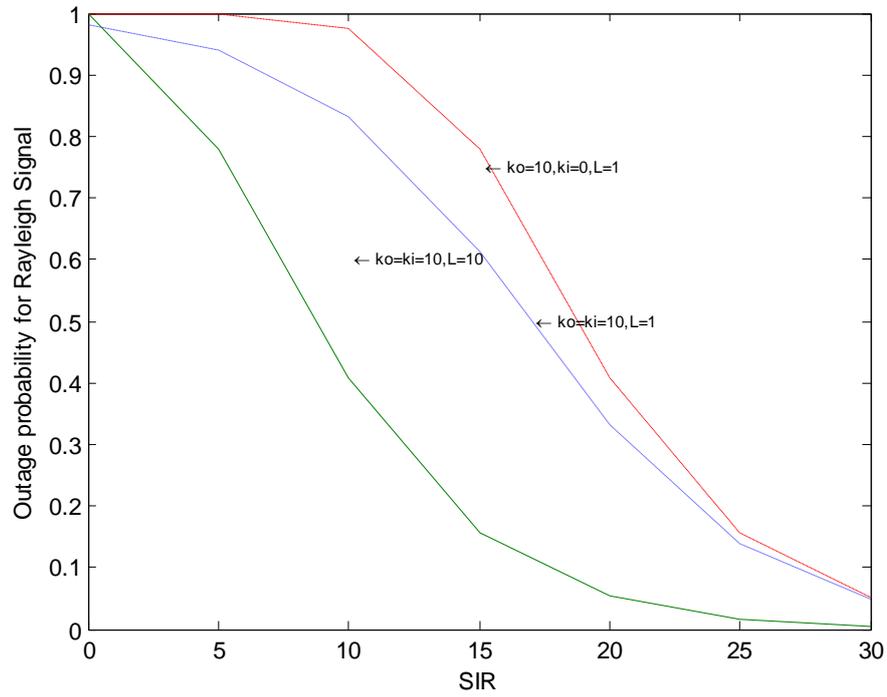


Fig. 1. Outage probability versus SIR for various L, K_0 , and K_1 , With $R_1 = 5$

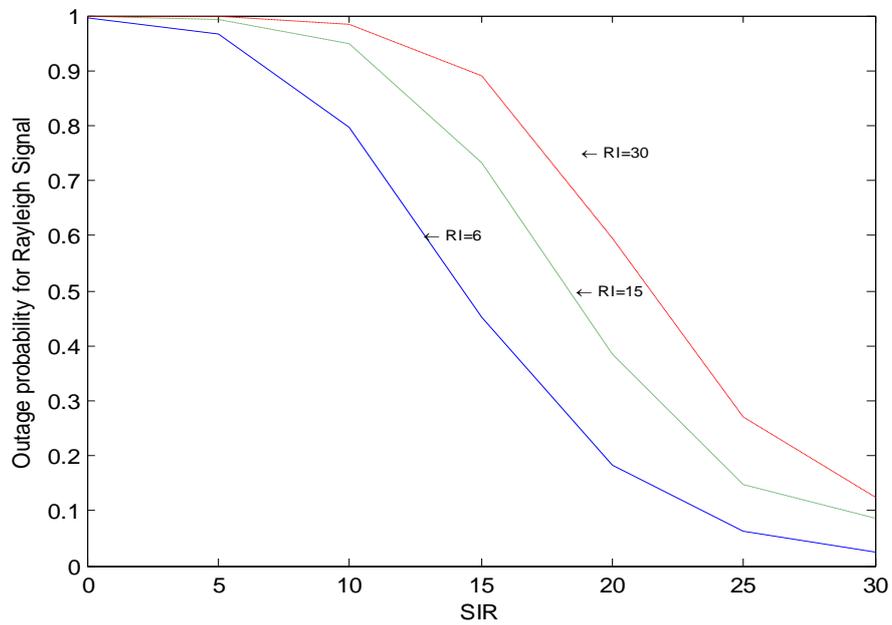


Fig.2. Outage probability versus SIR for various R_1 , with $K_0 = 10$ and $K_1 = 5$ for desired Rayleigh signal

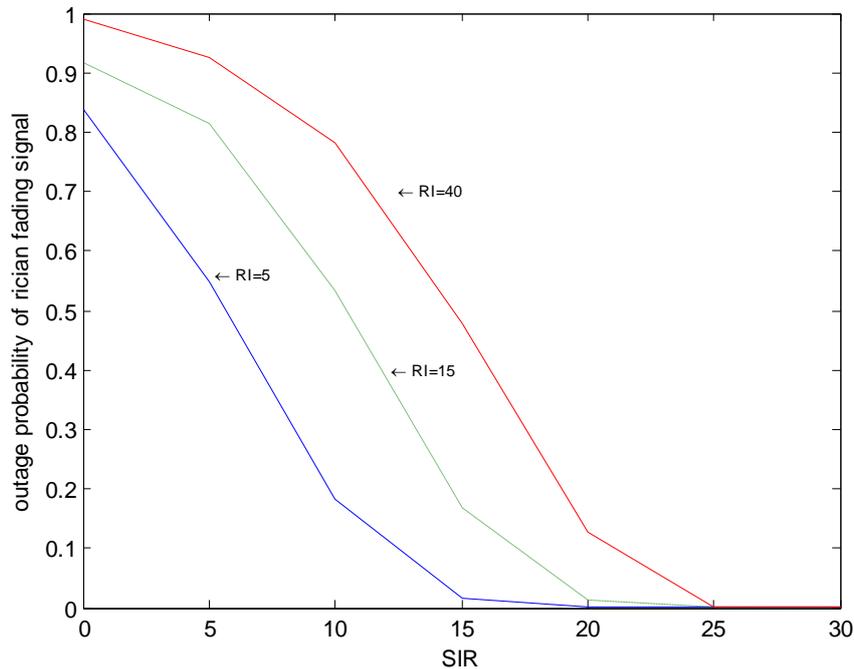


Fig.3. Outage probability versus SIR for various and R_i , with $K_0 = 10$ and $K_1 = 0$ for desired Rician signal

V. CONCLUSIONS

In this paper, a novel exact derivation has been presented to compute outage probabilities due to interference in mobile radio channels with Rayleigh/ Rician fading, We have also presented different graph form of outage probability for the Rician and Rayleigh fading environment without considering any type of shadowing effect on microcellular mobile radio systems. The effect of Signal to Interference ratio (SIR), Protection ratio (R_i), Rice factor of signals K , and number of cochannel interferes L on the outage probability have been investigated. New Rician/Rayleigh outage probabilities graphs for different values of protection ratios are presented and discussed. The results are almost same as previous results by this technique.

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